

# Designing & Implementation of Digital Filter for removal of Power Supply Noise

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**Abstract:** An ECG is a simple and useful test which records the rhythm and electrical activity of the heart of the patient that suffers from any heart disease. An ECG can detect problems you may have with your heart rhythm. It can help doctors tell if you are having a heart attack or if you've had a heart attack in the past. Sometimes an ECG can indicate if your heart is enlarged or thickened.

Digital Filter Design problem involves the determination of a set of filter coefficients to meet a set design specifications. These specifications typically consist of the width of the pass band and the corresponding gain, the width of the stop band(s) and the attenuation therein; the band edge frequencies (which give an indication of the transition band) and the peak ripple tolerable in the pass band and stop band(s). There are many techniques for selecting coefficients. We can use a spreadsheet like Microsoft Excel, or there are many design packages which will do the job. I have used MATLAB for this purpose as it is the most advanced tool for DSP applications. Also it helps to verify the design and results that comes from the hardware.

**Index Terms:** FIR, IIR, Matlab, VHDL.

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## I. INTRODUCTION

There are various methods to remove the noise of the ECG signal which may involve the IIR or FIR filter each has its own advantages and disadvantages. FIR filter because of its finite impulse response is always stable but its number of coefficients is very large, so it needs a larger memory space to store its coefficients. On the other hand the IIR filter has less number of coefficients and can be unstable sometimes due feedback loop involved in it. Essentially, Equation for FIR filtering is a 1-D convolution between the filter coefficients and the input data. In performing convolution, one of the two sets of numbers is reversed and "slid past" the other.

The resulting stream of numbers is found by taking the sum of the multiplications at each sliding interval. Like the IIR structure, the FIR realization can be highly replica table, which becomes important in the hardware design. One important aspect of FIRs is the linear phase characteristic, which makes it ideal for most digital signal processing applications [1,2]. Non recursive filters are always stable unlike the recursive or IIR filters which have to keep the pole placements in perspective. Again, FIRs have to have twice the order of an IIR because they cannot achieve the smaller side lobes in the stop band of the frequency response given the same number of parameters as an IIR. Despite the higher order of the FIR filter, the implementation is feasible in hardware and possesses the necessary linear phase property needed by channel models. Filter properties, design criteria, and the application at hand determine from which filter to choose.

## II. CHARACTERISTICS OF IIR FILTERS

1. Filter output depends upon previous inputs, present inputs and also on previous outputs.
2. IIR filters are useful for high-speed designs because they typically require a lower number of multiply compared to FIR filters.
3. IIR filters can be designed to have a frequency response that is a discrete version of the

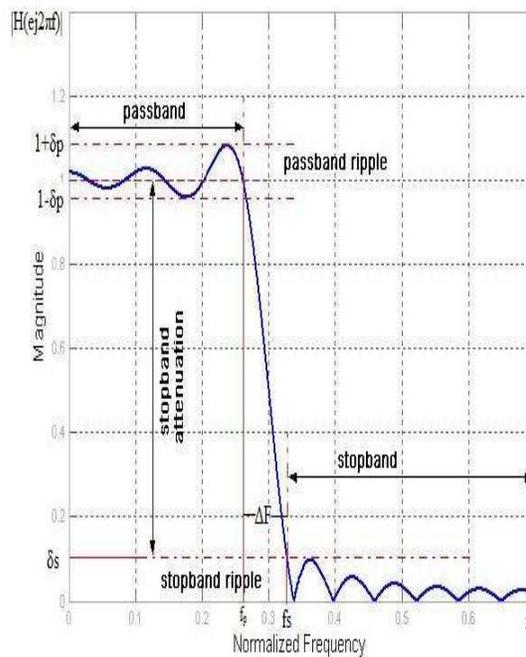


and results that comes from the hardware. I method, the frequency sampling method an

niques for IIR filters include the windowing max (optimal filter) method.

### III. DESIGNING OF DIGITAL IIR FILTER USING MATLAB & MODELSIM

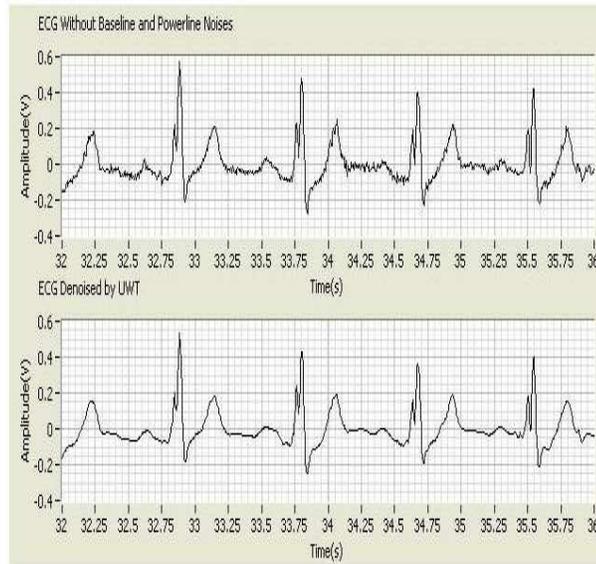
In the project I have used the Parks-McClellan Algorithm to calculate the coefficients. This algorithm gives the minimum order Filter for the given specifications. The filter thus designed is Equiripple Linear phase IIR filter. In MATLAB rerecord and remez function's are used to calculate minimum order and coefficients of filter. To calculate each output of the FIR filter, we multiply a set of samples by a set of coefficients. When a new sample arrives, it is added to the sample set, and the oldest sample is disposed of. This can be performed using the circular addressing hardware [5,6,7].



#### Steps Involved in Designing IIR Filter

1. Filter Specification
2. Find filter coefficients that meet the specification
3. Calculation of Coefficient quantization and quantization noise effects
4. Realization of filter structure (Direct Form, Transposed Form, Cascade, or Lattice)

The computational algorithm implementing equation of an IIR filter can be conveniently represented in block diagram. It is done using building blocks elements such as Multipliers, Adders and the Unit Delays. This way of presenting the difference equations in the form of block diagram and Signal Flow Diagram makes easy to write an algorithm, which can be implemented in the digital computer. As mentioned above Digital Filter Design problem involves the determination of a set of filter coefficients to meet a set of design specifications. A typical ECG waveform has been shown in fig.



These specifications typically consist of the width of the pass band and the corresponding gain, the width of the stop band(s) and the attenuation therein; the band edge frequencies (which give an indication of the transition band) and the peak ripple tolerable in the pass band and stop band(s). Many signal processing tasks frequently necessitate an immense amount of floating-point or fixed-point calculations for real-time or near real-time speeds. Such tasks require special-purpose signal processing architectures using specialized data paths, optimized sequencing, and pipelining. Unfortunately; such systems forego much flexibility despite operating at sufficient speeds. Digital finite impulse response filtering introduces one of many computationally demanding signal processing tasks. Wireless indoor channel modeling can be represented by an IIR filter using complex arithmetic due to the magnitude and phase responses of the channel impulse characteristics [7,8] We have done Comparison of IIR filters for Removal of ECG Baseline drift by using various structures.

IIR filter	Order	Spectral density before filtration (dB/Hz)	Spectral density after filtration (dB/Hz)	Average Power of signal (dB)	Ringing Effect
Butterworth	2	40.74	32.77	39.48	Large
Chebyshev Type I	2	40.74	31.39	38.97	Large
Chebyshev Type II	2	40.74	27.23	39.87	Small
Elliptic	2	40.74	31.13	39.00	Large

Although not new to the realm of programmable devices, field programmable gate arrays (FPGAs) are becoming increasingly popular for rapid prototyping of designs with the aid of software simulation and synthesis. Software synthesis tools translate high-level language descriptions of the implementation into formats that may be loaded directly into the FPGAs. An increasing number of design changes through software synthesis become more cost effective than similar changes done for hardware prototypes. In addition, the implementation may be constructed on existing hardware to help further reduce the cost [8, 9].

When considering trade-offs between hardware resources and performance, it can be useful to evaluate the effects of filter taps and finite precision arithmetic on the filter response. However, some of the following plots can be misleading. Finite precision arithmetic places a noise floor on plots which can easily be misinterpreted as a floor of the filter response. The Comparison of the complexity of different IIR filters has been shown in table.

Structure	Number of multiplications	Number of additions and subtractions	Total number of operations	Required bitwidth
Direct form	16	16	40	21
Cascade	13	16	34	12
Parallel	18	16	39	11
Continued fraction	18	16	35	23
Ladder	17	32	50	14
Wave digital	11	30	47	12

However, it is possible to replace a general purpose DSP chip and design special hardware digital filters which will operate at video-speed sampling rates. In other cases, the speed limitations can be overcome by first storing the high speed ADC data in a buffer memory. The buffer memory is then read at a rate which is compatible with the speed of the DSP-based digital filter. A comparison also can be done between IIR and FIR on basis of adders and multipliers shown in the following table:

Filter	Order of filter	No. of adders	No. of multipliers	No. of delays
FIR	450	450	451	450
IIR	2	4	4	2

In this manner, pseudo-real-time operation can be maintained as in a radar system, where signal processing is typically done on bursts of data collected after each transmitted pulse. Even in highly over-sampled sampled data systems, an analog anti-aliasing filter is still required ahead of the ADC and a reconstruction (anti-imaging) filter after the DAC. Finally, as signal frequencies increase sufficiently, they surpass the capabilities of available ADCs, and digital filtering then becomes impossible. Active analog filtering is not possible at extremely high frequencies because of op-amp bandwidth and distortion limitations, and filtering requirements must then be met using purely passive components [9, 10].



The input wave consisting of the two waves has been shown in fig below

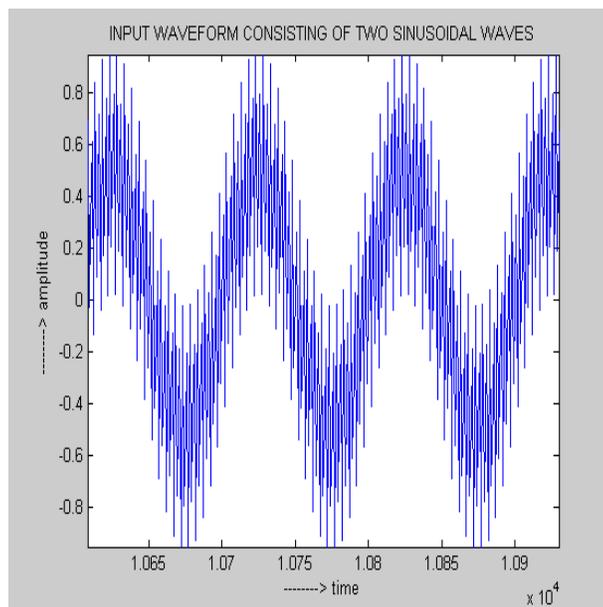


Figure2: Input wave consisting of two sine waves

When input wave has been applied to digital filter the output consists of one wave form which is desired wave and all other waves has been removed by the filter. As shown below out of two input waves only required wave is obtained in output. Basically all other waves are Noises which are added in ECG signal .

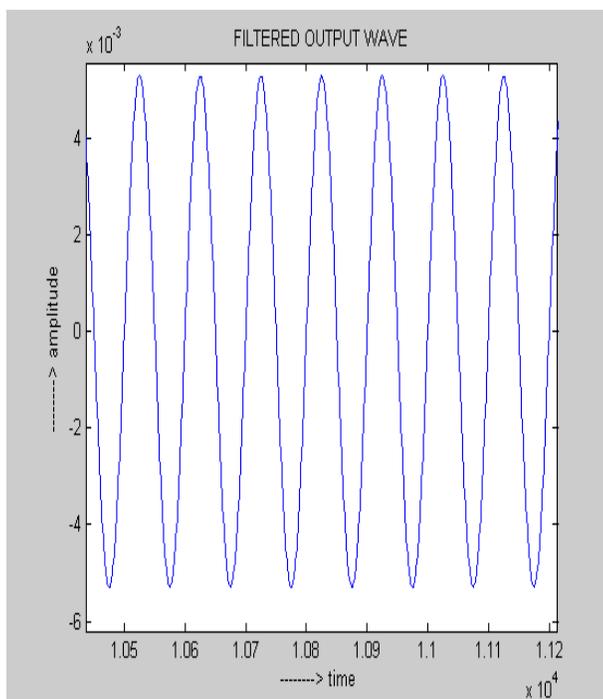


Figure 3: Output of the filter when the input of figure 2 was given as input

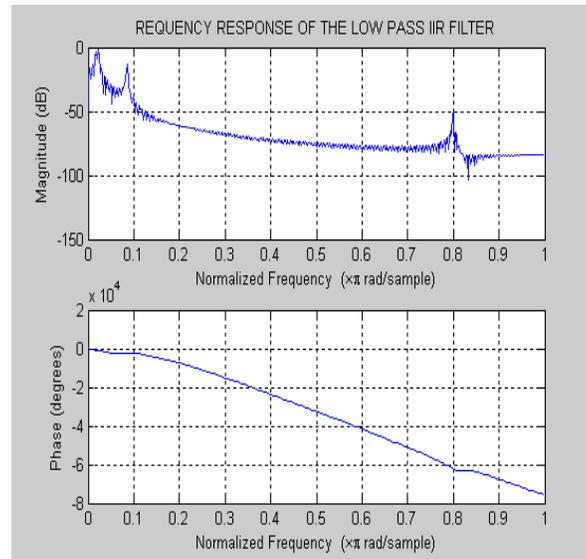


Figure 4: Frequency response (magnitude and phase ) of filter

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